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# **Diversity Gain Influenced by Polarization and Spatial Diversity Techniques in Ultrawideband**

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**ABSTRACT** This paper presents an experimental investigation of diversity gain influenced by polarization and spatial diversity techniques in ultrawideband (UWB) radio technology. To this aim, two different coplanar-fed UWB diversity antennas having the same size and employing identical monopoles were taken and both effective and apparent diversity gain were measured in a reverberation chamber. To extract the diversity gain, three commonly used approaches to combine diversity (i.e., selection, equal gain, and maximal ratio) have been considered. Results showed that >1 dB ( $\sim$ 26%) improvement in diversity gain is obtained over most of the considered band for polarization diversity case as compared with spatial case, showing the usefulness of polarization diversity for future UWB diversity applications.

**INDEX TERMS** Diversity gain, polarization and spatial, UWB diversity antenna, reverberation chamber.

# I. INTRODUCTION

Diversity proved to be an effective solution to mitigate multipath fading signals and enhance the system capacity. The most popular diversity technique is the antenna diversity which can be classified as spatial/space, polarization and pattern diversities [1], [2]. It is widely used in body-centric wireless communications due to offering significant improvements for instance for the on-body channels [3], [4]. Moreover, it can benefit from ultra-wideband (UWB) radio technology due to its large bandwidth which improves the number of channels allowing higher data rates and resolution (see for instance [5]). This also facilitates the compatibility of such technology in body-centric applications. There are some studies presented in the literature investigating the benefits of employing spatial or polarization diversity techniques. For instance, in [6] the spatial diversity has been investigated as it doesn't need any additional spectrum whereas in [7] with the motivation of having high isolation considering the smaller space needed and the reduced cost for installation the polarization diversity was preferred.

Diversity gain (DG) is a measure of reliability, estimating the decays of error probability with the increase of signal-to-noise ratio (SNR) at a given outage probability. Recently, another term, *i.e.* effective DG, has been introduced as an absolute value of DG enabling the comparison between different diversity antennas [8]. In classical narrow-band applications, spatial diversity is a common

method to combat fading [9]. The study presented in [10] has shown that, under the considered conditions, the DG brought by spatial diversity is about 0.7 dB more important than the one brought by polarization diversity. Moreover, results in [7] have shown that the aperture-coupled patch provides at least 1.5 dB more DG compared to a slanted dipole configuration. Regarding the DG in UWB case, the studies in [6] and [11] investigate the DG for on- and off-body communications with the antenna employing spatial diversity. In [12], an experimental study of both spatial and polarization diversity with multiple-input-multipleoutput (MIMO) antennas has been presented for body-centric applications. However, no DG discussion and result were provided. To the authors' knowledge the influence of UWB spatial and polarization diversities on DG has been unreported.

In this paper, an investigation of DG influenced by both spatial and polarization diversities has been performed in the frame of a specific UWB scenario. For this purpose, two UWB diversity antennas having the same size and employing identical monopoles were taken and both effective and apparent DGs were measured in a reverberation chamber.

#### **II. DIVERSITY GAIN CONCEPTS**

Diversity gain definition is conditioned by the probability that the SNR is above a reference level.

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Mathematically, its general expression is [13]

$$DG = \left[\frac{\gamma_c}{\Gamma_c} - \frac{\gamma_1}{\Gamma_1}\right]$$

where  $\gamma_c$  is the instantaneous SNR of the diversity combined signal,  $\Gamma_c$  is the mean SNR of the combined signal,  $\gamma_1$  is the highest SNR of the diversity branch signals, and  $\Gamma_1$  is the mean value of  $\gamma_1$ .

The received power level in a multipath environment with no line of sight is statistically distributed as a Rayleigh function [8]. Moreover, in a diversity scheme with two branches, the received power from each of them will have a Rayleigh-shaped probability density function. For this, the probability that an arbitrary power level sample is smaller than a certain power level; that is, the cumulative distribution function (CDF) is plotted. To this aim, the squared power samples  $(|S_{21}|^2)$  are simply sorted and then the series on the x axis against i/N on the y axis are plotted (where i is the index of the sorted series ranging from 1 to N).

Theoretically, the difference between the strongest channel trace and the combined signal represents the DG for each different combining technique, for a fixed level of the CDF of the received signals [14]. In order to optimize the SNR in a diversity scheme, the output signals can be combined in several ways such as selection combining (SC), equal gain combining (EGC), and maximal ratio combining (MRC) [15].

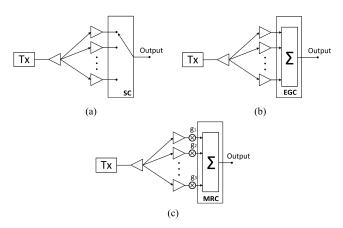


FIGURE 1. Three common methods of combining signals: (a) SC; (b) EGC; and (c) MRC.

#### A. SELECTION COMBINING

It selects the strongest received signal in branches [16], [17]. Fig.1a shows the SC block-diagram. The key idea is that, at any given moment of time, it picks up the branch with the highest SNR. Note that to provide same SNR for each branch, a number of modulators are required to adjust the gain of each diversity branch.

# B. EQUAL GAIN COMBINING

It gain all received signals with the same values and coherently combines all [16], [17]. Fig.1b shows the EGC block diagram. The key idea is that it linearly combines

the branch signals which leads to a high SNR output stronger than each branch signal. Note that all signals are co-phased to provide equal gain combining.

#### C. MAXIMAL RATIO COMBINING

It gains more in better branches and gains less in bad branches [16], [17]. Fig.1c shows the MRC block diagram. The key idea is that it uses linear coherent combining of the branch signals which leads to the maximized SNR output. Note that all signals must be co-phased before being summed. This requires a phasing circuit for each antennas element.

If the two received levels are combined according to a certain diversity combination rule, the diversity gain relative to the reference branch, can then be expressed as [8]:

$$DG = \frac{P_{div}}{P_{branch}}$$

where  $P_{div}$  is the power level after diversity combining, and  $P_{branch}$  is the power level of the reference branch. The two power levels must be read at the same cumulative probability level, which normally is taken to be 0.01; that is, 1%. It should be noted that the cumulative power distribution for the combined case will be located to the right of the curves for the two branches. DG is a useful tool for evaluating the diversity performance, but it might not express the reality. In reality, the antenna performance encounters different losses, accumulated in the antenna radiation efficiency. Recently, another term, *i.e.* effective DG, has been introduced which is an absolute measure of DG [8]. It also enables comparing the DG of different diversity antennas. It can be mathematically expressed as:

$$DG_{effective} = \left[\frac{P_{div}}{P_{branch}}\right]. \quad \eta_{radiation} = \left[\frac{P_{div}}{P_{ideal}}\right]$$

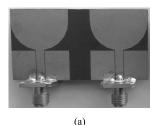
where  $\eta_{radiation}$  is the radiation efficiency of the reference branch, and  $P_{ideal}$  is the received power level of a single antenna with unit radiation efficiency and located in the same environment.  $P_{div}$  and  $P_{branch}$  must also be measured at the same cumulative probability levels. In a CDF plot, the effective diversity gain can be seen as the difference along the abscissa axis between the ideal reference and the diversity curve at some specific outage probability.

# **III. ANTENNA STRUCTURE AND DESIGN**

To be able to study the spatial and polarization effects on DG and to have a fair comparison, we have designed a new dual-port coplanar waveguide-fed (CPW) UWB diversity antenna having the same size  $(27 \times 52 \times 1.57 \text{ mm}^3)$  and identical monopoles as the polarization diversity antenna reported in [18] but employing spatial diversity. In this case, the space between the two branches is 8 mm. The detailed design of the polarization diversity antenna was presented in [18]. Photographs of both prototypes are shown in Fig. 2. For the rest of this paper and to simplify the descriptions, the spatial and the polarization UWB diversity antennas will be referred to as SP and POL, respectively.

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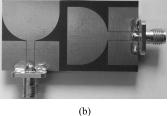


FIGURE 2. Photographs of UWB diversity antennas: (a) SP; (b) POL [18].

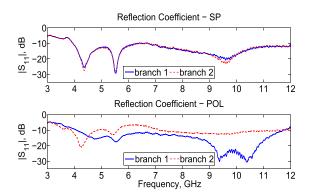
**TABLE 1.** Measurement setup information.

Measurement technique	Reverberation chamber
Measurement type	chamber calibration (e.g. REF antenna), reflection coefficient, total efficiency, diversity gain
Chamber loading	AUT with 50-Ω termination and REF antenna
REF antenna	BLUETEST discone REF antenna S/N: 001-C-005
Instrument used	Agilent E8363B, Agilent ECAL
Freq. range	3.0 GHz – 12 GHz (with steps of 6 MHz)
Freq. stirring	24 MHz
Stirring sequence	3 Stirrers - stepped sequence mode: 200 mechanical stirrer positions and 3 fixed antennas inside the chamber.
No. of diversity antennas	2
No. of samples per freq.	600
No. of freq. points	1501

# **IV. MEASUREMENT SETUP**

Multiple-mode fields inside reverberation chambers can be expanded in terms of eight times more plane waves than excited modes [19]. It emulates a rich multipath propagation environment [20], and can be used to accurately measure both embedded radiation efficiencies [21] and DGs [22], [23]. The DGs can also easily be estimated from the embedded element efficiencies and correlation between ports using simple equations [24]. In this study, the whole measurements were conducted in the reverberation chamber at Chalmers University, with dimensions  $1.9 m \times 2.0 m \times 1.4 m$ , one of the chambers characterized in [21]. The detailed measurement setup information is given in Table 1. Note that measurements to evaluate the DG have been carried out from 3 to 12 GHz; performance at higher frequencies was measured considering the limitation in the measurement equipment, e.g. the reference (REF) antenna. As seen in Table 1, the REF antenna is a standard discone antenna developed by BLUETEST [25]. It is well matched (based on  $|S_{11}| \le -10$  dB) from 0.7-9.5 GHz and its impedance up to 12 GHz is below -6 dB. The same setup was used for the calibrations. During the measurements, when one branch is connected to the network analyzer, the other branch is terminated with a  $50-\Omega$ load and vice versa. In order to obtain the DG, at first the power samples of the two branches of each diversity antennas are combined using the three common methods,

*i.e.* SC, EGC and MRC. Then, the CDF of each branch's power and the combined power samples need to be evaluated. Finally, by estimating the average power level difference between the CDFs, at the certain cumulative probability level (normally 1%), of the combined case and the strongest branch the DG is obtained.



**FIGURE 3.** Measured antennas'  $|S_{11}|$  in reverberation chamber.

# **V. EXPERIMENTAL RESULTS AND DISCUSSION**

Fig. 3 illustrates measured  $|S_{II}|$  of both prototypes in the reverberation chamber demonstrating their ultra-wide matching bandwidth. For both antennas the  $|S_{II}|$  was checked in an anechoic chamber as well; for the POL, looking at the results presented in [18] we note that the matching levels are very similar when measured in a reverberation chamber. The isolation between the two branches of the SP was also checked and found to be higher than 15 dB across the entire band of interest.

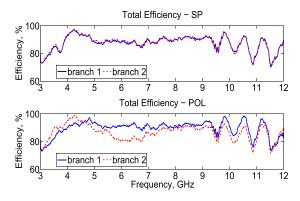


FIGURE 4. Measured antennas' total efficiency in reverberation chamber.

In order to be able to compare the DG of both antennas, the effective DG must be calculated [8]. For this, total efficiency of both antennas is required. Fig. 4 depicts the measured total efficiency of the considered antennas. As observed, when both antennas are matched (4-12 GHz), an average total efficiency of almost 90% for SP and of about 80%-91% for POL is obtained for each of their branches. The fluctuations at higher frequencies 9.5-12 GHz is due to the matching issues of the used reference antenna, which has a similar behavior.

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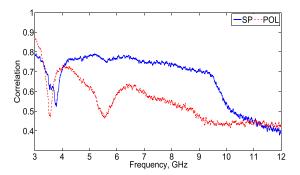


FIGURE 5. Measured correlation between antenna ports for both SP and POL.

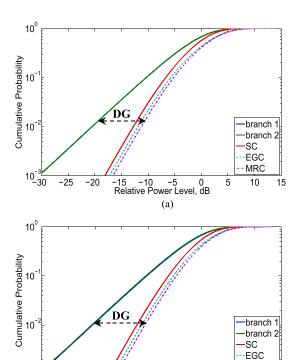


FIGURE 6. CDF plot, showing diversity gain at 1% outage probability, calculated from reverberation chamber measurement: (a) SP; (b) POL.

Relative Power Level, dB

(b)

10\_-30

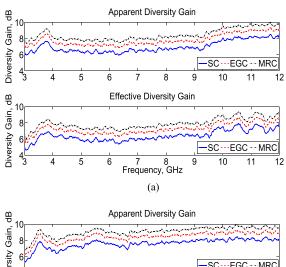
-25 -20

- MRC

10

15

The apparent DG performance for SC, EGC and MRC in reverberation chamber is related to efficiency and correlation [26]. The correlation between antenna ports for SP and POL cases is plotted in Fig. 5. As seen, the correlation in the interval where the reference antenna is well matched is higher in the SP than the POL case which makes POL diversity perform better and provide improved DG results. To confirm this, the received power at each antenna's branches was combined through SC, EGC, and MRC techniques and the CDFs were calculated; Fig. 6 shows the results. As observed, by taking the power difference of the strongest branch, here both have similar power levels, and the combined case the DG is obtained. Fig. 7 illustrates both apparent and effective



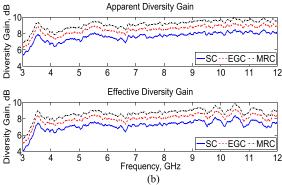


FIGURE 7. Measured apparent and effective diversity gains of the two antennas assuming 1% outage probability: (a) SP; (b) POL.

DGs of both antennas achieved with the considered combining techniques at outage probability of 1%. As seen, in all cases the DGs slightly improve with the increase of the frequency. The effective DG values, considering the frequency interval where both antennas have more stable efficiencies 4-9.5 GHz, for the SP/SC is about 6 dB whereas for the POL/SC it is about 7 dB. The MRC technique results in a higher effective DG for the both considered antennas; about 7.5 dB and 8.5 dB for SP and POL, respectively. This shows that both diversity antennas are suitable for UWB diversity applications, however, the POL leads to at least 1 dB (~ 26%) improvement in DG. The fluctuations appearing at the higher frequencies of the effective DG are due to the relation with the antenna's total efficiency. The repeatability of the measurements was checked, and similar results obtained.

### VI. CONCLUSION

An experimental comparison of diversity gain between polarization and spatial diversity techniques in UWB systems has been made. In order to have a fair comparison, two different dual-port CPW-fed UWB diversity antennas having the same size and identical monopoles were taken as representative antennas. The diversity gain was measured in a reverberation chamber which provides characteristics similar to a multipath propagation environment. Three common diversity combining techniques

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were considered. Results showed a diversity gain at 1% cumulative probability around 7.5 and 8.5 dB for SP/MRC and POL/MRC, respectively. Both antennas ensure good diversity behavior; however, the POL improves the diversity gain of about 1 dB.

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