Random Sampling of Bandlimited Signals on Graphs

Pierre Vandergheynst

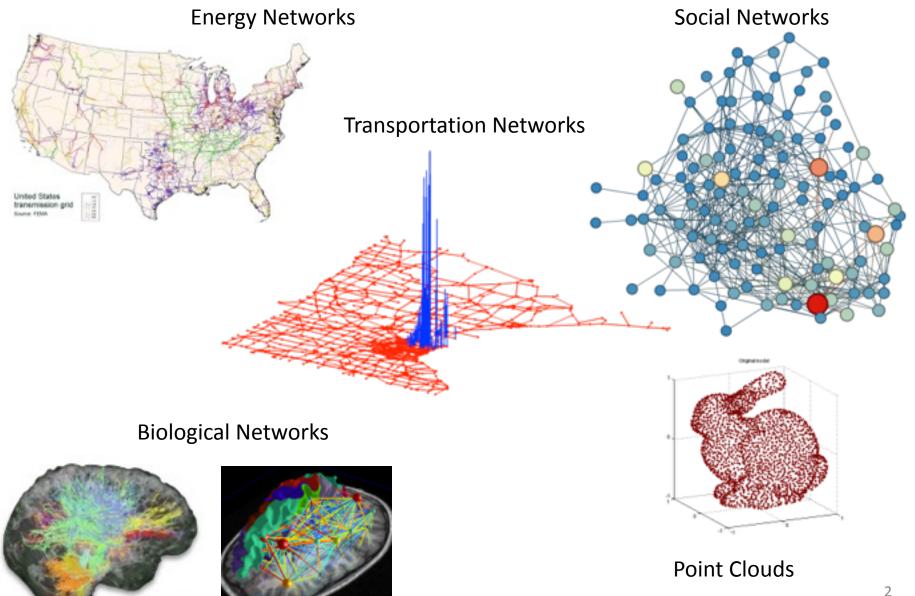
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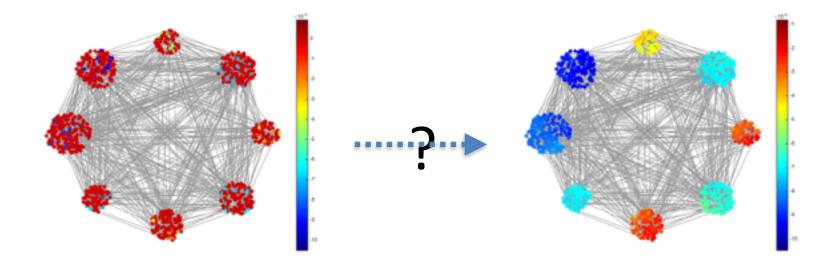
NIPS2015 Workshop Multiresolution Methods for Large Scale Learning

Motivation



Goal

Given partially observed information at the nodes of a graph



Can we robustly and efficiently infer missing information?

What signal model?

How many observations?

Influence of the structure of the graph?

Notations

$$\mathcal{G} = \{\mathcal{V}, \mathcal{E}, W\}$$
 weighted, undirected

 \mathcal{V} is the set of n nodes

 \mathcal{E} is the set of edges

 $W \in \mathbb{R}^{n \times n}$ is the weighted adjacency matrix

$$L \in \mathbb{R}^{n \times n}$$

combinatorial graph Laplacian L := D - W

normalised Laplacian $L := I - D^{-1/2}WD^{-1/2}$

diagonal degree matrix D has entries $d_i := \sum_{i \neq j} W_{ij}$

Notations

L is real, symmetric PSD

orthonormal eigenvectors $\mathsf{U} \in \mathbb{R}^{n \times n}$ Graph Fourier Matrix non-negative eigenvalues $\pmb{\lambda}_1 \leqslant \pmb{\lambda}_2 \leqslant \dots, \pmb{\lambda}_n$

$$L = U\Lambda U^{\mathsf{T}}$$

Notations

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non-negative eigenvalues $\lambda_1 \leqslant \lambda_2 \leqslant \ldots, \lambda_n$

$$L = U\Lambda U^{\mathsf{T}}$$

k-bandlimited signals $x \in \mathbb{R}^n$

Fourier coefficients $\hat{x} = \mathsf{U}^\intercal x$

$$oldsymbol{x} = \mathsf{U}_k \hat{oldsymbol{x}}^k \qquad \hat{oldsymbol{x}}^k \in \mathbb{R}^k$$

 $\mathsf{U}_k := (oldsymbol{u}_1, \dots, oldsymbol{u}_k) \in \mathbb{R}^{n imes k}$ first k eigenvectors only

$$p \in \mathbb{R}^n$$
 $p_i > 0$ $||p||_1 = \sum_{i=1}^n p_i = 1$ $P := \operatorname{diag}(p) \in \mathbb{R}^{n \times n}$

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Draw independently m samples (random sampling)

$$\mathbb{P}(\omega_j = i) = \mathbf{p}_i, \quad \forall j \in \{1, \dots, m\} \text{ and } \forall i \in \{1, \dots, n\}$$

$$m{p} \in \mathbb{R}^n$$
 $m{p}_i > 0$ $m{\|p\|}_1 = \sum_{i=1}^n m{p}_i = 1$ $P := \mathrm{diag}(m{p}) \in \mathbb{R}^{n \times n}$

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$$\mathbb{P}(\omega_j = i) = \mathbf{p}_i, \quad \forall j \in \{1, \dots, m\} \text{ and } \forall i \in \{1, \dots, n\}$$

$$oldsymbol{y}_j := oldsymbol{x}_{\omega_j}, \quad orall j \in \{1, \dots, m\}$$
 $oldsymbol{y} = \mathsf{M} oldsymbol{x}$

$$\frac{\left\|\mathsf{U}_{k}^{\mathsf{T}}\boldsymbol{\delta}_{i}\right\|_{2}}{\left\|\mathsf{U}^{\mathsf{T}}\boldsymbol{\delta}_{i}\right\|_{2}} = \frac{\left\|\mathsf{U}_{k}^{\mathsf{T}}\boldsymbol{\delta}_{i}\right\|_{2}}{\left\|\boldsymbol{\delta}_{i}\right\|_{2}} = \left\|\mathsf{U}_{k}^{\mathsf{T}}\boldsymbol{\delta}_{i}\right\|_{2}$$

How much a perfect impulse can be concentrated on first k eigenvectors

Carries interesting information about the graph

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Graph Coherence

$$\nu_{\boldsymbol{p}}^{k} := \max_{1 \leqslant i \leqslant n} \left\{ \boldsymbol{p}_{i}^{-1/2} \left\| \mathbf{U}_{k}^{\mathsf{T}} \boldsymbol{\delta}_{i} \right\|_{2} \right\}$$
Rem: $\nu_{\boldsymbol{p}}^{k} \geqslant \sqrt{k}$

Theorem 1 (Restricted isometry property). Let M be a random subsampling matrix with the sampling distribution \mathbf{p} . For any $\delta, \epsilon \in (0,1)$, with probability at least $1 - \epsilon$,

$$(1 - \delta) \|\boldsymbol{x}\|_{2}^{2} \leqslant \frac{1}{m} \|\mathsf{MP}^{-1/2} \boldsymbol{x}\|_{2}^{2} \leqslant (1 + \delta) \|\boldsymbol{x}\|_{2}^{2}$$
 (1)

for all $\mathbf{x} \in \text{span}(\mathsf{U}_k)$ provided that

$$m \geqslant \frac{3}{\delta^2} (\nu_{\boldsymbol{p}}^k)^2 \log \left(\frac{2k}{\epsilon}\right).$$
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Proof similar to CS in bounded ONB but simpler since model is a subspace (not a union)

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Can we reduce to optimal amount?

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Variable Density Sampling
$$p_i^* := \frac{\|\mathbf{U}_k^{\mathsf{T}} \boldsymbol{\delta}_i\|_2^2}{k}, \quad i = 1, \dots, n$$
 is such that: $(\nu_{\boldsymbol{p}}^k)^2 = k$ and depends on structure of graph

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Corollary 1. Let M be a random subsampling matrix constructed with the sampling distribution \mathbf{p}^* . For any $\delta, \epsilon \in (0,1)$, with probability at least $1-\epsilon$,

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for all $\mathbf{x} \in \text{span}(\mathsf{U}_k)$ provided that

$$m \geqslant \frac{3}{\delta^2} k \log \left(\frac{2k}{\epsilon}\right).$$

$$oldsymbol{y} = \mathsf{M} oldsymbol{x} + oldsymbol{n} \qquad oldsymbol{y} \in \mathbb{R}^m$$
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Standard Decoder

$$\min_{\boldsymbol{z} \in \operatorname{span}(\mathsf{U}_k)} \left\| \mathsf{P}_{\Omega}^{-1/2} \left(\mathsf{M} \boldsymbol{z} - \boldsymbol{y} \right) \right\|_2$$

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re-weighting for RIP

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Efficient Decoder:

$$\min_{\boldsymbol{z} \in \mathbb{R}^n} \left\| \mathsf{P}_{\Omega}^{-1/2} \left(\mathsf{M} \boldsymbol{z} - \boldsymbol{y} \right) \right\|_2^2 + \gamma \; \boldsymbol{z}^{\intercal} g(\mathsf{L}) \boldsymbol{z}$$

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soft constrain on frequencies efficient implementation

Standard Decoder:

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Theorem 1. Let Ω be a set of m indices selected independently from $\{1,\ldots,n\}$ with sampling distribution $\mathbf{p} \in \mathbb{R}^n$, and M the associated sampling matrix. Let $\epsilon, \delta \in (0,1)$ and $m \geqslant \frac{3}{\delta^2} (\nu_{\mathbf{p}}^k)^2 \log \left(\frac{2k}{\epsilon}\right)$. With probability at least $1 - \epsilon$, the following holds for all $\mathbf{x} \in \mathrm{span}(\mathsf{U}_k)$ and all $\mathbf{n} \in \mathbb{R}^m$.

i) Let x^* be the solution of Standard Decoder with y = Mx + n. Then,

$$\|\boldsymbol{x}^* - \boldsymbol{x}\|_2 \leqslant \frac{2}{\sqrt{m(1-\delta)}} \|\mathsf{P}_{\Omega}^{-1/2} \boldsymbol{n}\|_2.$$
 (1)

ii) There exist particular vectors $\mathbf{n}_0 \in \mathbb{R}^m$ such that the solution \mathbf{x}^* of Standard Decoder with $\mathbf{y} = \mathsf{M}\mathbf{x} + \mathbf{n}_0$ satisfies

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non-negative

Efficient Decoder:

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Filter reshapes Fourier coefficients

$$h: \mathbb{R} o \mathbb{R}$$
 $oldsymbol{x}_h := \mathsf{U} \operatorname{diag}(\hat{oldsymbol{h}}) \, \mathsf{U}^\intercal oldsymbol{x} \in \mathbb{R}^n$ $\hat{oldsymbol{h}} = (h(oldsymbol{\lambda}_1), \dots, h(oldsymbol{\lambda}_n))^\intercal \in \mathbb{R}^n$

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$$p(t) = \sum_{i=0}^d lpha_i \, t^i$$
 $oldsymbol{x}_p = \mathsf{U} \, \mathrm{diag}(\hat{oldsymbol{p}}) \, \mathsf{U}^\intercal oldsymbol{x} = \sum_{i=0}^d lpha_i \, \mathsf{L}^i oldsymbol{x}$

Pick special polynomials and use e.g. recurrence relations for fast filtering (with sparse matrix-vector multiply only)

Efficient Decoder:

$$\min_{m{z} \in \mathbb{R}^n} \left\| \mathsf{P}_{\Omega}^{-1/2} \left(\mathsf{M} m{z} - m{y} \right) \right\|_2^2 + \gamma \, m{z}^\intercal g(\mathsf{L}) m{z}$$
 non-negative non-negative penalizes high-frequencies

Efficient Decoder:

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$$\text{non-decreasing} =$$

$$\text{penalizes high-frequencies}$$

Favours reconstruction of approximately band-limited signals

Ideal filter yields Standard Decoder

$$i_{\lambda_k}(t) := \begin{cases} 0 & \text{if } t \in [0, \lambda_k], \\ +\infty & \text{otherwise,} \end{cases}$$

Analysis of Efficient Decoder

Theorem 1. Let Ω , M, P, m as before and $M_{\max} > 0$ be a constant such that $\|MP^{-1/2}\|_2 \leq M_{\max}$. Let $\epsilon, \delta \in (0,1)$. With probability at least $1 - \epsilon$, the following holds for all $\mathbf{x} \in \text{span}(\mathsf{U}_k)$, all $\mathbf{n} \in \mathbb{R}^n$, all $\gamma > 0$, and all nonnegative and nondecreasing polynomial functions g such that $g(\boldsymbol{\lambda}_{k+1}) > 0$.

Let x^* be the solution of Efficient Decoder with y = Mx + n. Then,

$$\|\boldsymbol{\alpha}^* - \boldsymbol{x}\|_{2} \leq \frac{1}{\sqrt{m(1-\delta)}} \left[\left(2 + \frac{M_{\max}}{\sqrt{\gamma g(\boldsymbol{\lambda}_{k+1})}} \right) \| \mathsf{P}_{\Omega}^{-1/2} \boldsymbol{n} \|_{2} + \left(M_{\max} \sqrt{\frac{g(\boldsymbol{\lambda}_{k})}{g(\boldsymbol{\lambda}_{k+1})}} + \sqrt{\gamma g(\boldsymbol{\lambda}_{k})} \right) \|\boldsymbol{x}\|_{2} \right],$$

$$(1)$$

and

$$\|\boldsymbol{\beta}^*\|_2 \leqslant \frac{1}{\sqrt{\gamma g(\boldsymbol{\lambda}_{k+1})}} \|\mathsf{P}_{\Omega}^{-1/2} \boldsymbol{n}\|_2 + \sqrt{\frac{g(\boldsymbol{\lambda}_k)}{g(\boldsymbol{\lambda}_{k+1})}} \|\boldsymbol{x}\|_2,$$
 (2)

where $\boldsymbol{\alpha}^* := \mathsf{U}_k \mathsf{U}_k^{\mathsf{T}} \boldsymbol{x}^*$ and $\boldsymbol{\beta}^* := (\mathsf{I} - \mathsf{U}_k \mathsf{U}_k^{\mathsf{T}}) \boldsymbol{x}^*$.

Analysis of Efficient Decoder

Noiseless case:

$$\|\boldsymbol{x}^* - \boldsymbol{x}\|_2 \leqslant \frac{1}{\sqrt{m(1-\delta)}} \left(M_{\max} \sqrt{\frac{g(\boldsymbol{\lambda}_k)}{g(\boldsymbol{\lambda}_{k+1})}} + \sqrt{\gamma g(\boldsymbol{\lambda}_k)} \right) \|\boldsymbol{x}\|_2 + \sqrt{\frac{g(\boldsymbol{\lambda}_k)}{g(\boldsymbol{\lambda}_{k+1})}} \|\boldsymbol{x}\|_2$$

 $g(\lambda_k) = 0$ + non-decreasing implies perfect reconstruction

Analysis of Efficient Decoder

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 $g(\lambda_k) = 0 + \text{non-decreasing implies perfect reconstruction}$

Otherwise:

choose γ as close as possible to 0 and seek to minimise the ratio $g(\lambda_k)/g(\lambda_{k+1})$

Choose filter to increase spectral gap?

Clusters are of course good

Noise:
$$\|\mathsf{P}_{\Omega}^{-1/2} \boldsymbol{n}\|_2 / \|\boldsymbol{x}\|_2$$

Need to estimate $\|\mathsf{U}_k^{\intercal} \boldsymbol{\delta}_i\|_2^2$

Need to estimate $\|\mathsf{U}_k^\intercal \boldsymbol{\delta}_i\|_2^2$

Filter random signals with ideal low-pass filter:

$$r_{b_{\lambda_k}} = \mathsf{U} \operatorname{diag}(\lambda_1, \dots, \lambda_k, 0, \dots, 0) \mathsf{U}^\intercal r = \mathsf{U}_k \mathsf{U}_k^\intercal r$$

$$\mathbb{E}\left(\boldsymbol{r}_{b_{\boldsymbol{\lambda}_{k}}}\right)_{i}^{2} = \boldsymbol{\delta}_{i}^{\intercal} \mathsf{U}_{k} \mathsf{U}_{k}^{\intercal} \mathbb{E}(\boldsymbol{r}\boldsymbol{r}^{\intercal}) \left. \mathsf{U}_{k} \mathsf{U}_{k}^{\intercal} \boldsymbol{\delta}_{i} \right. = \left. \left. \left\| \mathsf{U}_{k}^{\intercal} \boldsymbol{\delta}_{i} \right\|_{2}^{2} \right.$$

Need to estimate $\|\mathsf{U}_k^{\intercal} \boldsymbol{\delta}_i\|_2^2$

Filter random signals with ideal low-pass filter:

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$$\mathbb{E}\left(\boldsymbol{r}_{b_{\boldsymbol{\lambda}_{k}}}\right)_{i}^{2} = \boldsymbol{\delta}_{i}^{\mathsf{T}}\mathsf{U}_{k}\mathsf{U}_{k}^{\mathsf{T}}\,\mathbb{E}(\boldsymbol{r}\boldsymbol{r}^{\mathsf{T}})\,\mathsf{U}_{k}\mathsf{U}_{k}^{\mathsf{T}}\boldsymbol{\delta}_{i} = \left\|\mathsf{U}_{k}^{\mathsf{T}}\boldsymbol{\delta}_{i}\right\|_{2}^{2}$$

In practice, one may use a polynomial approximation of the ideal filter and:

$$ilde{m{p}_i} := rac{\sum_{l=1}^L {(m{r}_{c_{\lambda_k}}^l)_i^2}}{\sum_{i=1}^n \sum_{l=1}^L {(m{r}_{c_{\lambda_k}}^l)_i^2}}$$

$$L \geqslant \frac{C}{\delta^2} \log \left(\frac{2n}{\epsilon}\right)$$

Estimating the Eigengap

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Again, low-pass filtering random signals:

$$(1 - \delta) \sum_{i=1}^{n} \left\| \mathsf{U}_{j^*}^{\mathsf{T}} \boldsymbol{\delta}_i \right\|_2^2 \, \leqslant \, \sum_{i=1}^{n} \sum_{l=1}^{L} \, (\boldsymbol{r}_{b_{\lambda}}^l)_i^2 \, \leqslant \, (1 + \delta) \, \sum_{i=1}^{n} \left\| \mathsf{U}_{j^*}^{\mathsf{T}} \boldsymbol{\delta}_i \right\|_2^2$$

Estimating the Eigengap

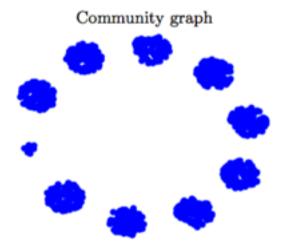
Again, low-pass filtering random signals:

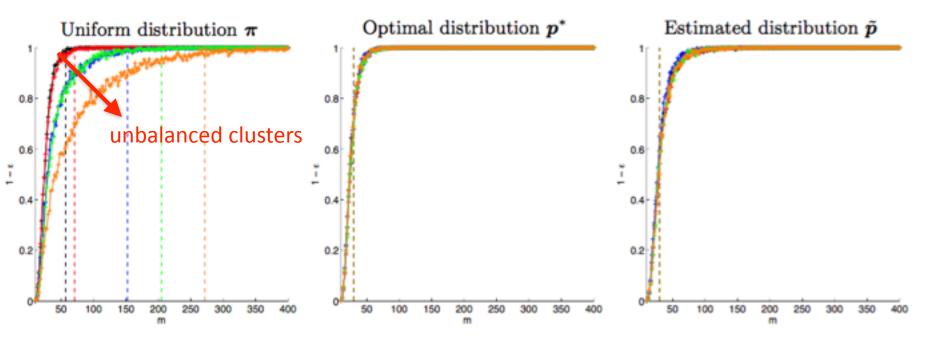
$$(1 - \delta) \sum_{i=1}^{n} \left\| \mathsf{U}_{j^*}^{\mathsf{T}} \boldsymbol{\delta}_i \right\|_2^2 \, \leqslant \, \sum_{i=1}^{n} \sum_{l=1}^{L} \, (\boldsymbol{r}_{b_{\lambda}}^l)_i^2 \, \leqslant \, (1 + \delta) \, \sum_{i=1}^{n} \left\| \mathsf{U}_{j^*}^{\mathsf{T}} \boldsymbol{\delta}_i \right\|_2^2$$

Since:
$$\sum_{i=1}^{n} \left\| \mathsf{U}_{j^*}^{\mathsf{T}} \boldsymbol{\delta}_i \right\|_2^2 = \left\| \mathsf{U}_{j^*} \right\|_{\text{Frob}}^2 = j^*$$

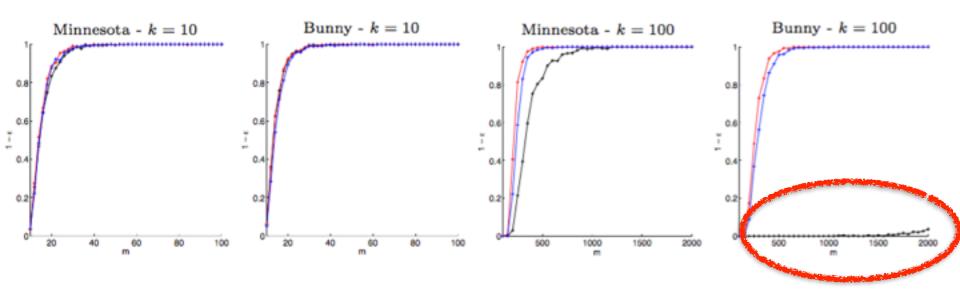
We have:
$$(1 - \delta) j^* \leq \sum_{i=1}^n \sum_{l=1}^L (r_{b_{\lambda}}^l)_i^2 \leq (1 + \delta) j^*$$

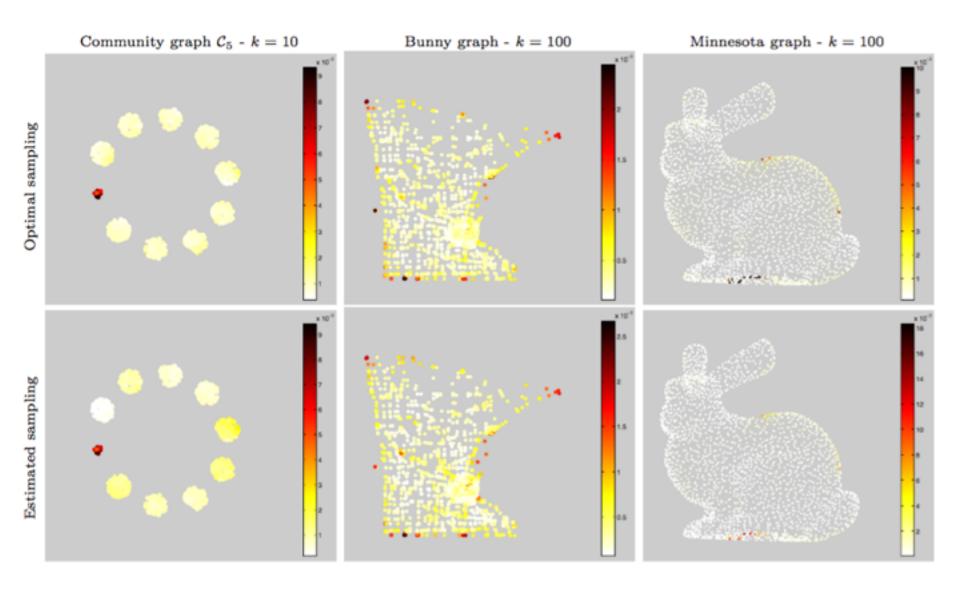
Dichotomy using the filter bandwidth

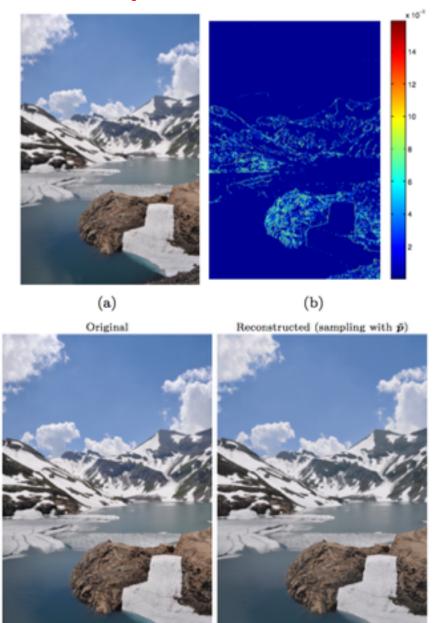












7%

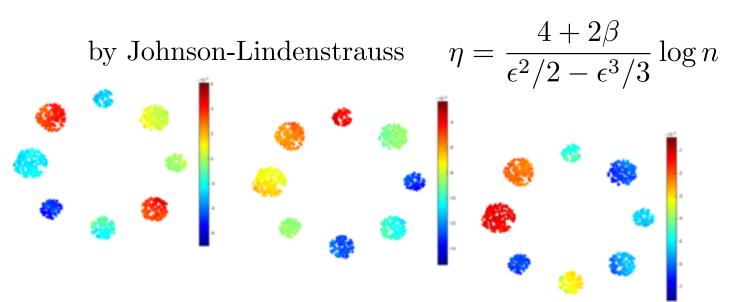
Clustering equivalent to recovery of cluster assignment functions

Well-defined clusters -> band-limited assignment functions!

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Generate features by filtering random signals



Clustering equivalent to recovery of cluster assignment functions

Well-defined clusters -> band-limited assignment functions!

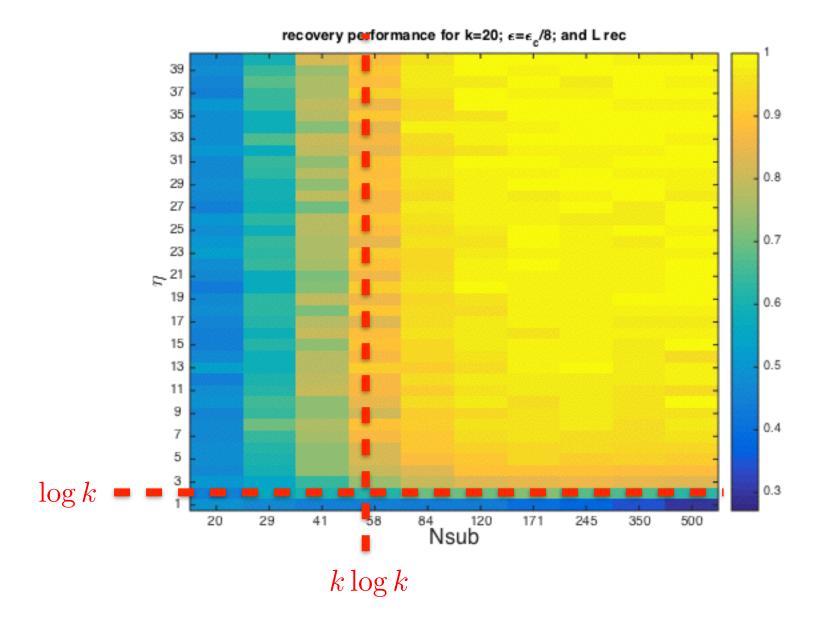
Generate features by filtering random signals

by Johnson-Lindenstrauss
$$\eta = \frac{4+2\beta}{\epsilon^2/2 - \epsilon^3/3} \log n$$

Each feature map is smooth, therefore keep

$$m \geqslant \frac{6}{\delta^2} \nu_k^2 \log\left(\frac{k}{\epsilon'}\right)$$

Use k-means on compressed data and feed into Efficient Decoder⁴



• **Stable, robust and universal random sampling** of smoothly varying information on graphs.

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- Tractable decoder with guarantees
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- Can be used for inference, (SVD less) compressive clustering

Thank you!